

# Modeling Synchronous Logic Circuits

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# Basic Sequential Circuits

- A combinational circuit produces output solely depending on the current input.
- But a sequential circuit “remembers” its previous state.
- Its output depends on present inputs and previous state.
- Examples:
  - Latches
  - Registers
  - Memory
  - parallel to serial / serial to parallel converters
  - Counters

# Latch vs Registers

- **Latch: Level sensitive device**
  - Positive Latches and Negative latches
  - Can be realized using multiplexers
- **Register: edge triggered storage element**
  - Can be implemented using latches
  - Cascade a negative latch with a positive latch to obtain a positive edge triggered register
- **Flip flop: bi-stable component formed by the cross coupling of gates.**

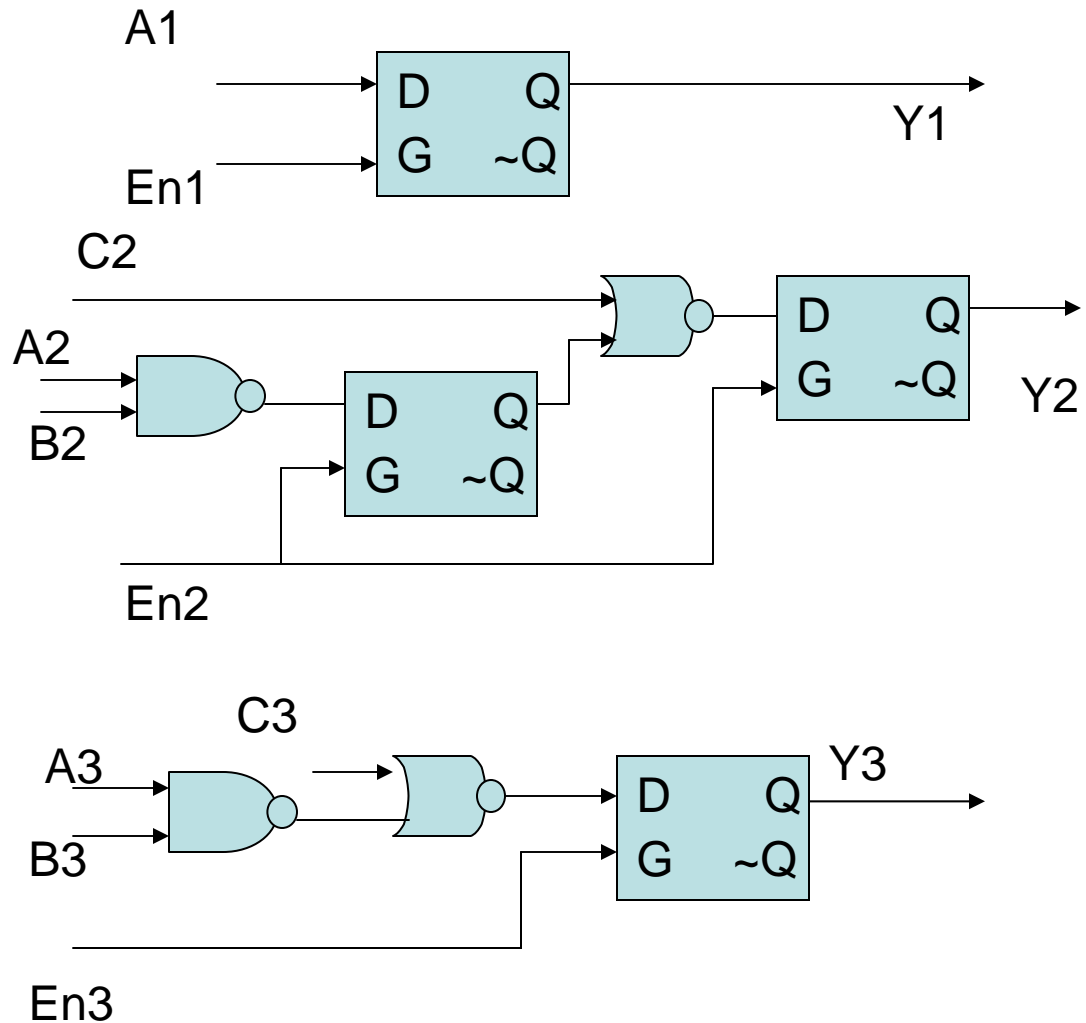
# Latches

- Cycle stealing is possible leading to faster circuits
- Problem of timing analysis.

# Latch inference using if

- ```

module ....
always@(...)
begin
    if(En1)
        Y1=A1;
    if(En2)
        begin
            M2<=!(A2&B2);
            Y2<=!(M2|C2);
        end
    if(En3)
        begin
            M3=!(A3&B3);
            Y3=!(M3|C3);
        end
end
        
```



# Modeling latches with present and clear inputs

- begin

- if(!Clear1)

- Y1=0;

- else if(En)

- Y1=A1;

- begin

- if(Clear2)

- Y2=0;

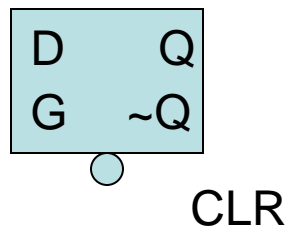
- else if(En)

- Y2=A2;

# Modeling latches with present and clear inputs

- if(!Preset3)  
    Y3=1;  
else if(En3)  
    Y3=A3;

- if(!Preset3)  
    Y3=1;  
else if(En3)  
    Y3=A3;



# Modeling latches with present and clear inputs

- if(Clear5)

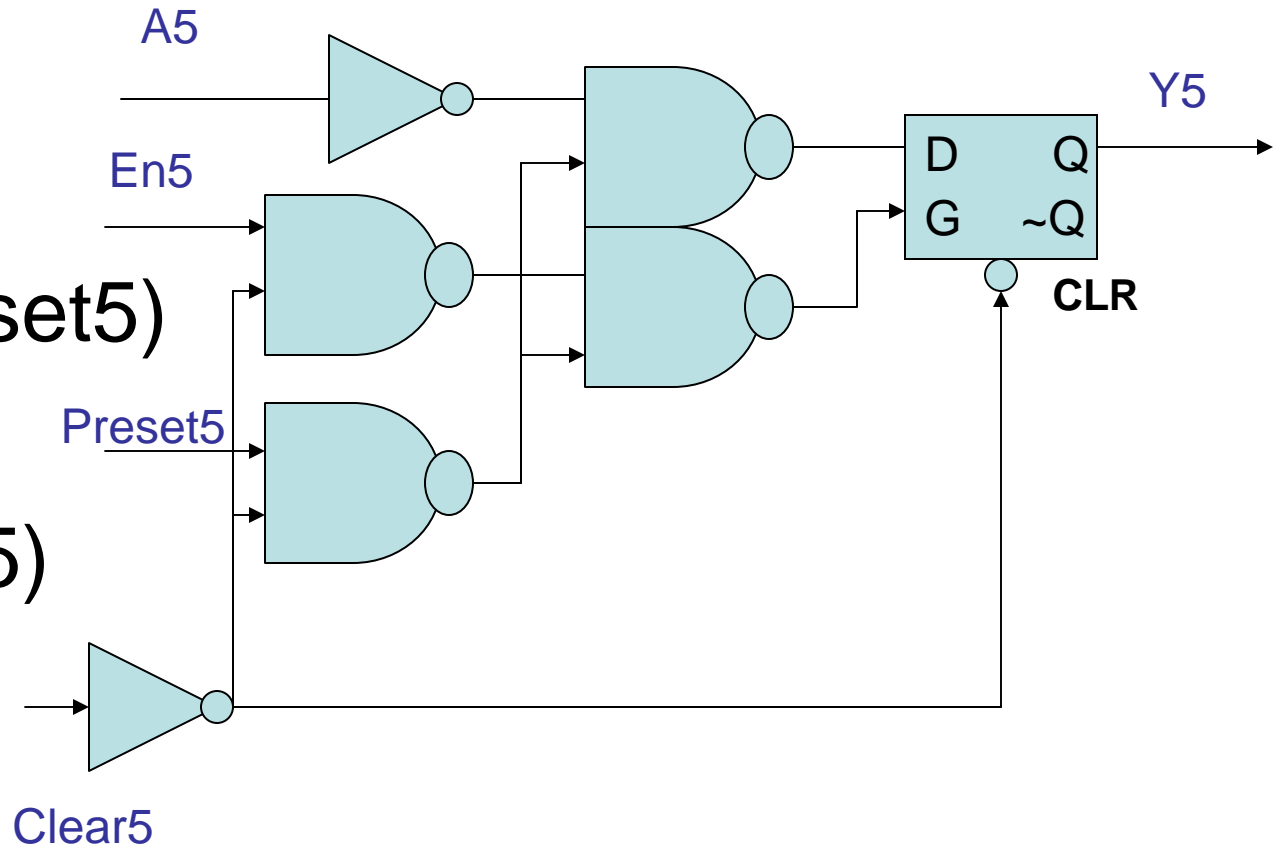
Y5=0;

else if(Preset5)

Y5=1;

else if(En5)

Y5=A5;



*If there are no latches with a preset input In the library, equivalent functionality is produced by using latches with a clear input.*



# Multiple gated latch

always @(En1 or En2 or En3 ...)

if(En1==1)

Y=A1;

else if(En2==1)

Y=A2;

else if(En3==1)

Y=A3;

Try to synthesize and check whether:

1. Is there a latch inferred?
2. Put an else statement. Is a latch inferred now?
3. Put a default output assignment before the if starts. Is a latch inferred now?
4. Use the posedge keyword in the trigger list, and repeat the above experiments.

# Other places of latch inferences

- Nested if: If all the possibilities are not mentioned in the code.
- Case: In advertent. Not advisable to infer a latch from case statement.
  - may lead to superfluous latches.
- Nested case statements can also infer latches.

# The D-Flip Flop

- always @(posedge clk)  
    Y=D;
- A-Synchronous reset:  
    always @(posedge clk or posedged reset)  
        if(reset)  
            Y=0  
        else Y=D;

# Resets

- Synchronous reset:  
always @(posedge clk)  
    if(reset)  
        Y=0  
    else Y=D;

# Combinational Block between two flops

- always@(posedge clk)

begin

M <= !(A & B);

Y <= M|N;

end

assign N=C|D;

*What will happen if a blocking assignment is used?*

**The first flip flop will become redundant...**

# Sequence Generators

- Linear Feedback Shift Registers
- Counters

# LFSR Applications

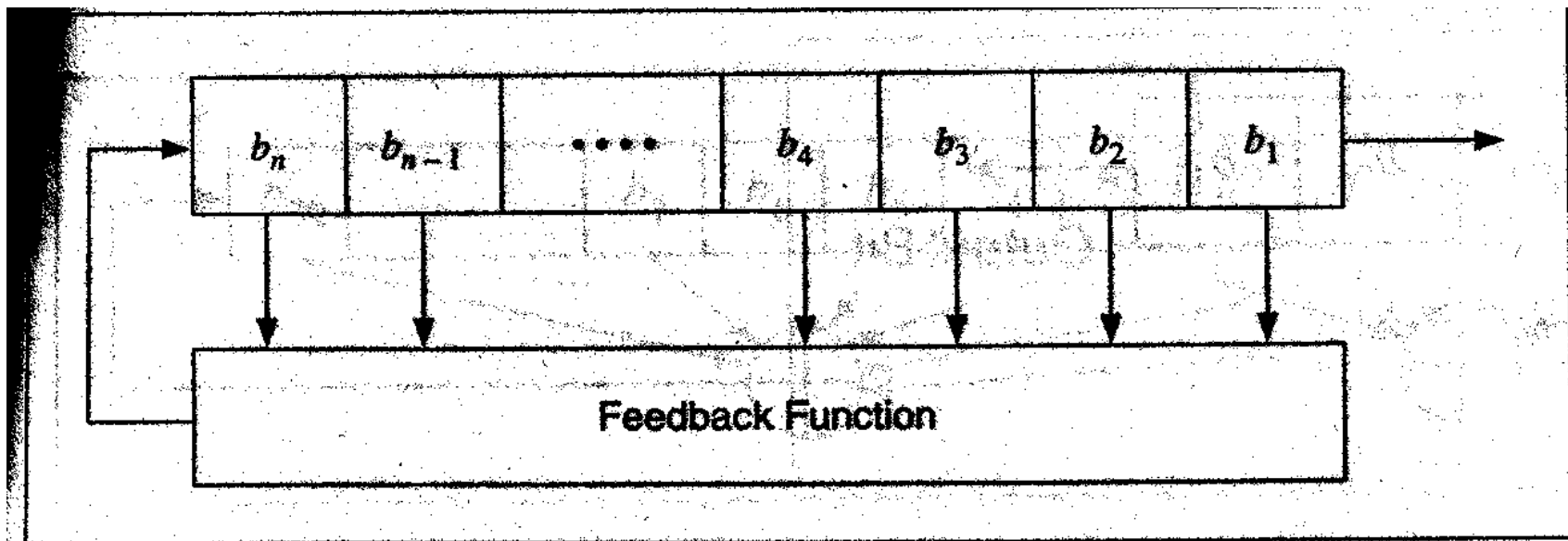
- Pattern Generators
- Counters
- Built-in Self-Test (BIST)
- Encryption
- Compression
- Checksums
- Pseudo-Random Bit Sequences (PRBS)

# LFSR

- **Linear Feedback Shift Register (LFSR):**
  - For pseudo random number generation
  - A shift register with feedback and exclusive-or gates in its feedback or shift path.
  - The initial content of the register is referred to as *seed*.
  - The position of XOR gates is determined by the polynomial (*poly*).

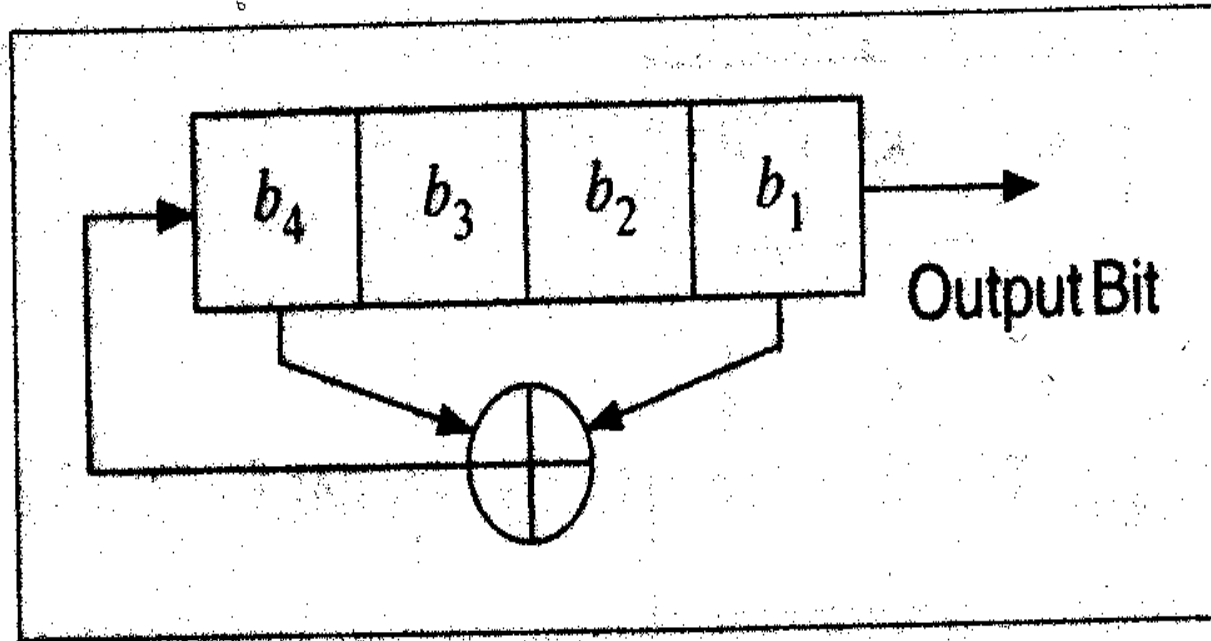


# An LFSR outline



*The feedback function (often called the taps) can be represented by a polynomial of degree  $n$*

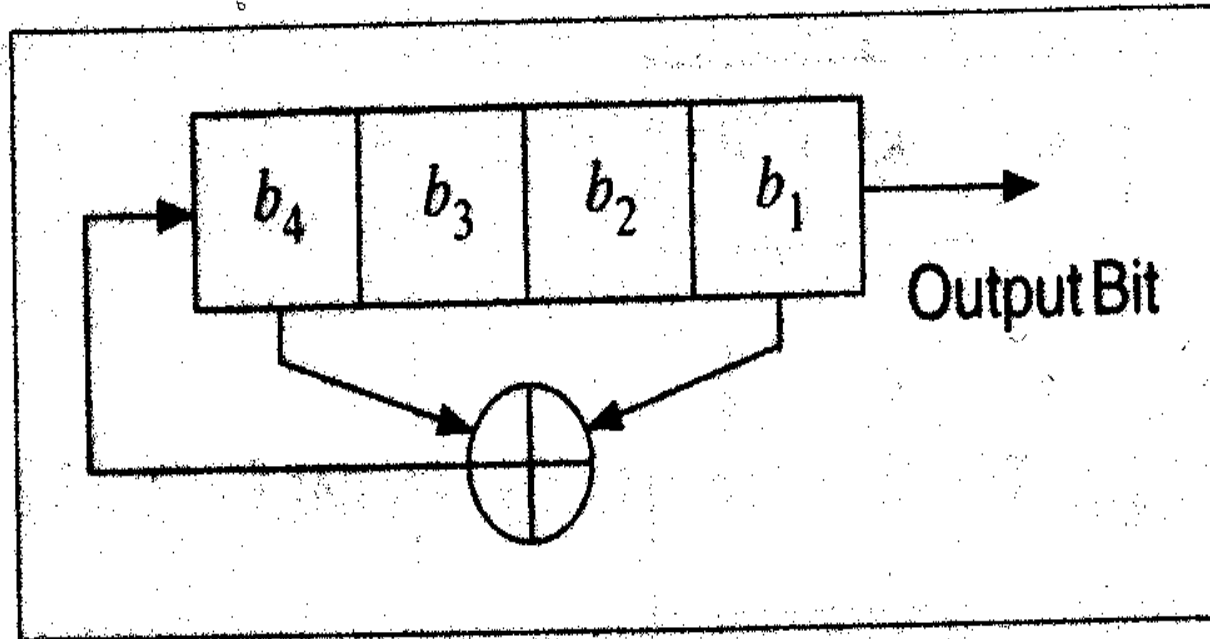
# A 4 bit LFSR



The feedback polynomial is  $p(x)=x^4+x+1$

# A 4 bit LFSR

1111  
0111  
1011  
0101  
1010  
1101  
0110  
0011  
1001  
0100  
0010  
0010  
1000  
1100  
1110



Output sequence:  
111101011001000...

All the  $2^4-1$  possible states are generated. This is called a maximal length LFSR. So, the sequence depends on the feedbacks.

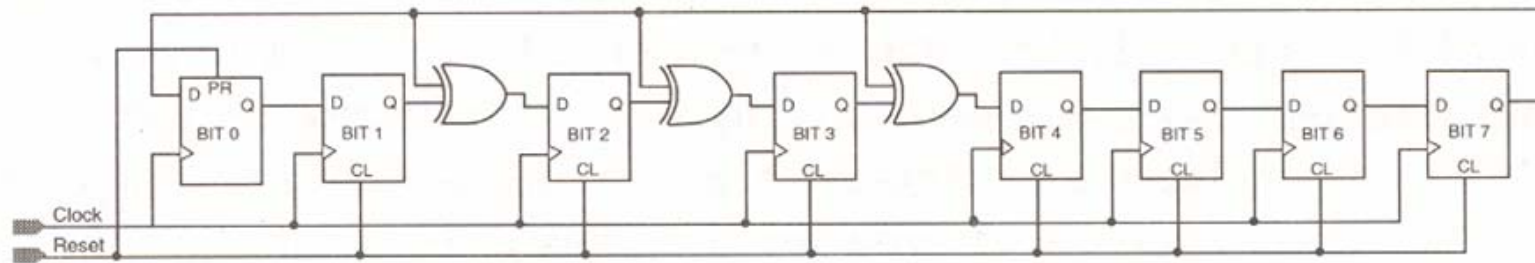
# Types of feedbacks

- Feedbacks can be comprising of XOR gates.
- Feedbacks can be comprising of XNOR gates.
- Given the same tap positions, both will generate the same number of values in a cycle. But the values will be same.
- Permutation!

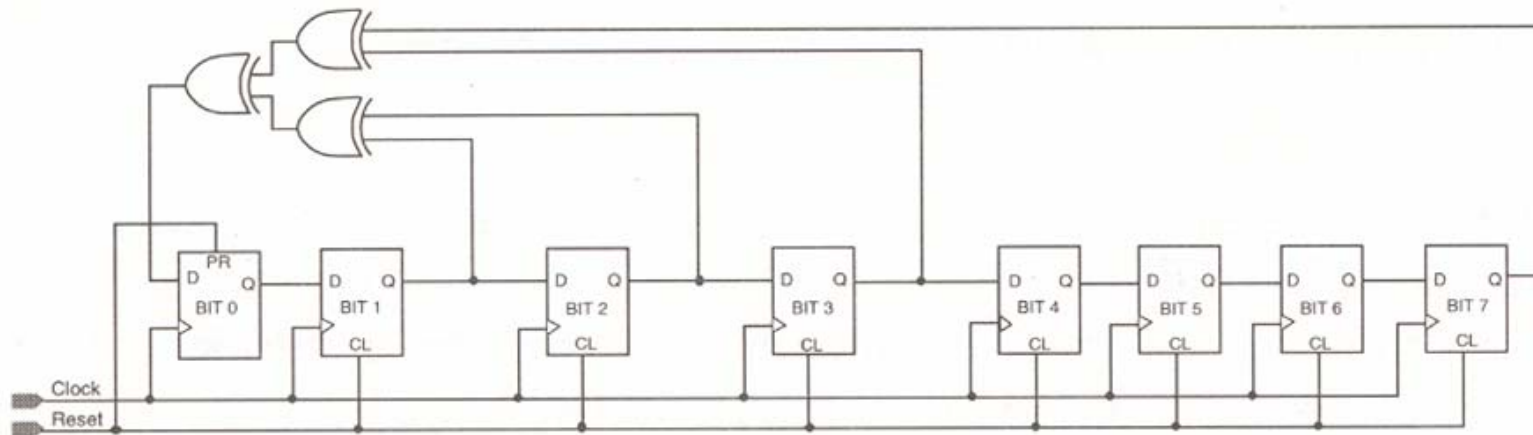
# Number of Taps

- For many registers of length  $n$ , only two taps are needed, and can be implemented with a single XOR (XNOR) gate.
- For some register lengths, for example 8, 16, and 32, four taps are needed. For some hardware architectures, this can be in the critical timing path.
- A table of taps for different register lengths is included in the back of this module.

# One-to-Many and Many-to-One



**(a) One-to-many**



Note. Uses XOR gates therefore all 0s not in sequence so not reset to all 0s.

**(b) Many-to-one**

Implementation (a) has only a single gate delay between flip-flops.

# Avoiding the Lockup State

## Will Use XOR Form For Examples

We have an n-bit LFSR, shifting to the “right”



# Avoiding the Lockup State

## Will Use XOR Form For Examples

The all '0's state can't be entered during normal operation but we can get close. Here's one of n examples:



We know this is a legal state since the only illegal state is all 0's. If the first  $n-1$  bits are '0', then bit 0 must be a '1'.



# Avoiding the Lockup State

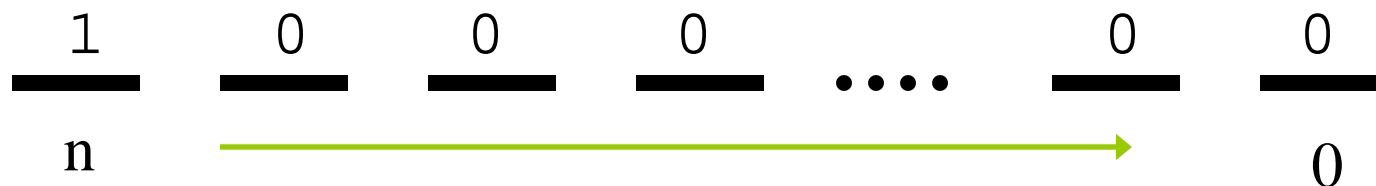
## Will Use XOR Form For Examples

Now, since the XOR inputs are a function of taps, including the bit 0 tap, we know what the output of the XOR tree will be: '1'.

It must be a '1' since '1' XOR '0' XOR '0' XOR '0' = '1'.



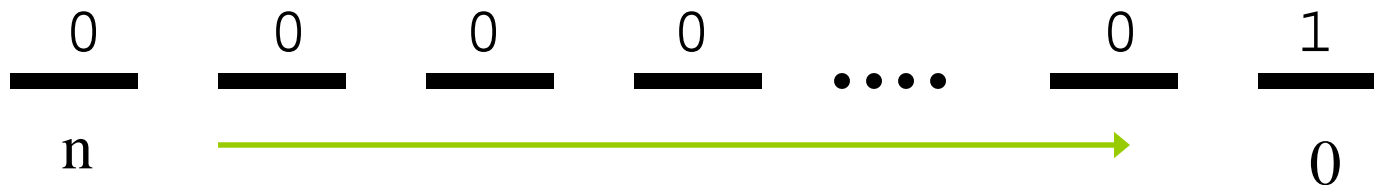
So normally the next state will be:



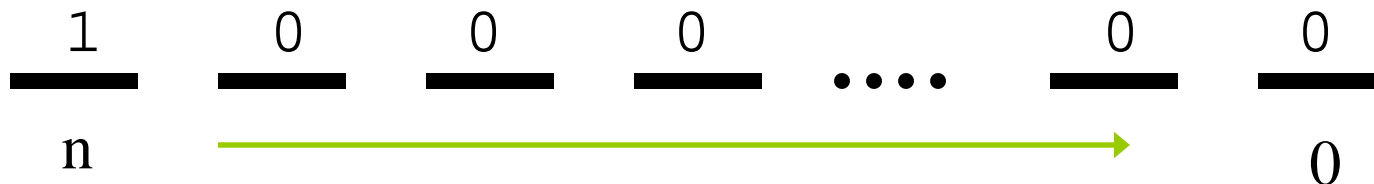
# Avoiding the Lockup State

## Will Use XOR Form For Examples

But instead, let's do this, go from this state:



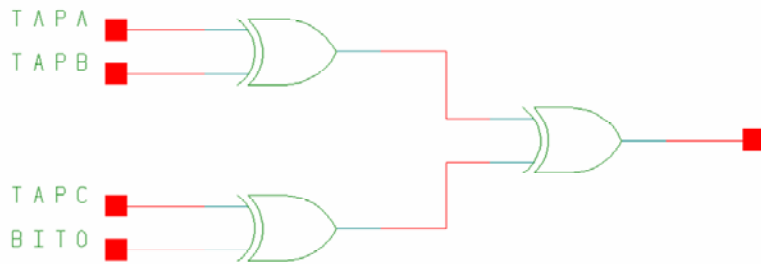
To the all '0's state:



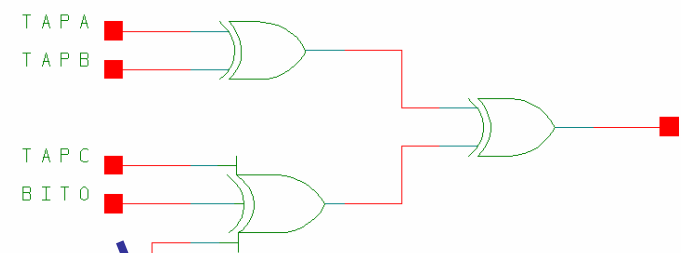
# Avoiding the Lockup State

## Modification to Circuit

$2^{n-1}$  states



$2^n$  states



NOR of all bits  
except bit 0



Added this term

- a) "000001" : 0 Xor 0 Xor 0 Xor 0 Xor 1 Xor 1  $\Rightarrow$  0
- b) "000000" : 0 Xor 0 Xor 0 Xor 0 Xor 0 Xor 1  $\Rightarrow$  1
- c) "100000" :

# Verilog code

```
module ...  
    always@(posedge clk or posedge rst)  
    begin  
        if(rst)  
            LFSR_reg=8'b0;  
        else  
            LFSR_reg=Next_LFSR_reg;  
        end
```

# verilog

```
always @(LFSR_reg)
begin
    Bits0_6_zero=~|LFSR_Reg[6:0];
    Feedback=LFSR_Reg[7]^ Bits0_6_zero;
    for(N=7;N>0;N=N-1)
        if(Taps[N-1]==1)
            Next_LFSR_Reg[N]=LFSR_Reg[N-1]^Feedback;
        else
            Next_LFSR_Reg[N]=LFSR_Reg[N-1];
    Next_LFSR_Reg[0]=Feedback;
end
assign Y=LFSR_Reg;
endmodule
```

# A Generic LFSR

```
module LFSR_Generic_MOD(Clk,rst,Y);  
  parameter width=8;  
  input clk,rst;  
  output [width-1:0] Y;  
  reg [31:0] Tapsarray [2:32];  
  wire [width-1:0] Taps;  
  integer N;  
  reg Bits0_Nminus1_zero, Feedback;  
  reg [width-1:0] LFSR_Reg, Next_LFSR_Reg;
```

```
always @(rst)
begin
    TapsArray[2]=2'b11;
    TapsArray[3]=3'b101;
    ...
    TapsArray[32]=32'b10000000_00000000_00000000_01100010;
end
assign Taps[width-1:0]=TapsArray[width];
```

**REST OF THE CODE IS SIMILAR TO THE PREVIOUS EXAMPLE**

# Counters

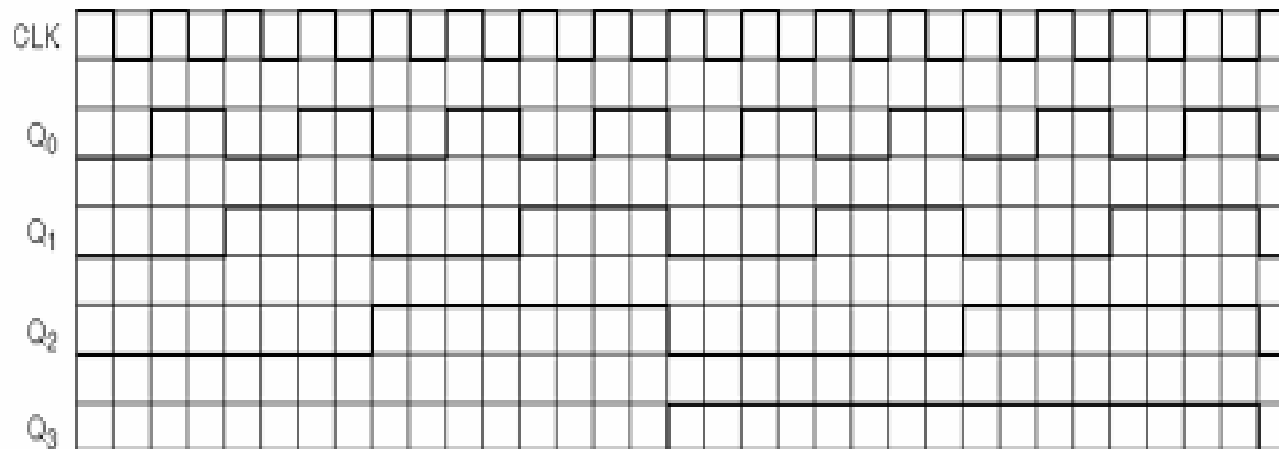
- A register that goes through a pre-determined sequence of binary values (states), upon the application of input pulses in one or more than inputs is called a counter.
- The input pulses can be random or periodic.
- Counters are often used as clock dividers.



# Timing Diagrams

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- The outputs ( $Q_0 \Rightarrow Q_3$ ) of the counter can be used as frequency dividers with  $Q_0 = \text{Clock} \div 2$ ,  $Q_1 = \text{Clock} \div 4$ ,  $Q_2 = \text{Clock} \div 8$ , and  $Q_3 = \text{Clock} \div 16$ .



# Types

- Synchronous
  - Using adders, subtractors
  - Using LFSRs, better performance because of simple circuits. Most feedback polynomials are trinomials or pentanomials.
- Asynchronous:
  - Ripple through flip flops
  - each single flip flop stage divides by 2
  - so, we may obtain division by  $2^n$
  - what if they are not powers of two? we require extra feedback logic
  - significantly smaller

# Divide by 13 :

## A synchronous design

```
always@(posedge clk or posedge rst)
```

```
begin
```

```
    if(!rst)
```

```
        begin
```

```
            cnt<=startcnt;
```

```
            Y<=0;
```

```
        end
```

# Divide by 13 :

## A synchronous design

else

if(Count==EndCount)

begin

Count<=StartCount;

Y<=1;

end

# Divide by 13 :

## A synchronous design

```
else
begin
  for(N=1;N<=3;N=N-1)
    if(Taps[N])
      Count[N]<=Count[N-1]~^Count[3];
    else
      Count[N]<=Count[N-1];
  Count[0]<=Count[3];
  Y=0;
end
end
```

# Asynchronous Design

- Instantiate 4 DFFs.
- Ripple the clock through them
- Output is a divided by 16 clock.
- Use the output states and check when 13 clock cycles have elapsed.
- Use it to make the output bit high.
- Reset the Flip Flops
- Exercise: Write a verilog code!

# Pros and Cons of Synchronous and Asynchronous Resets

# Problem of Choice

- Quite a complex issue.
- All of us know the importance of the reset button. When our PC does not work!
- Less understood, less emphasized.
- Require to a treatment to perform an informed design.



# Some Points

- Reset style depends on the ASIC design style, the application and where the flip flop is located.
- If we design considering all the unused states (like the  $2^n - 2n$  states in a Johnson's Counter), we should be able to do reset from any possible state.
- A power on reset is required if the designer used the unused states as don't cares to do optimization.
- Often an explicit reset is not required if the flop is a part of shift register, just wait for some clock cycles. These are often called **follower** flops

# Good Reset

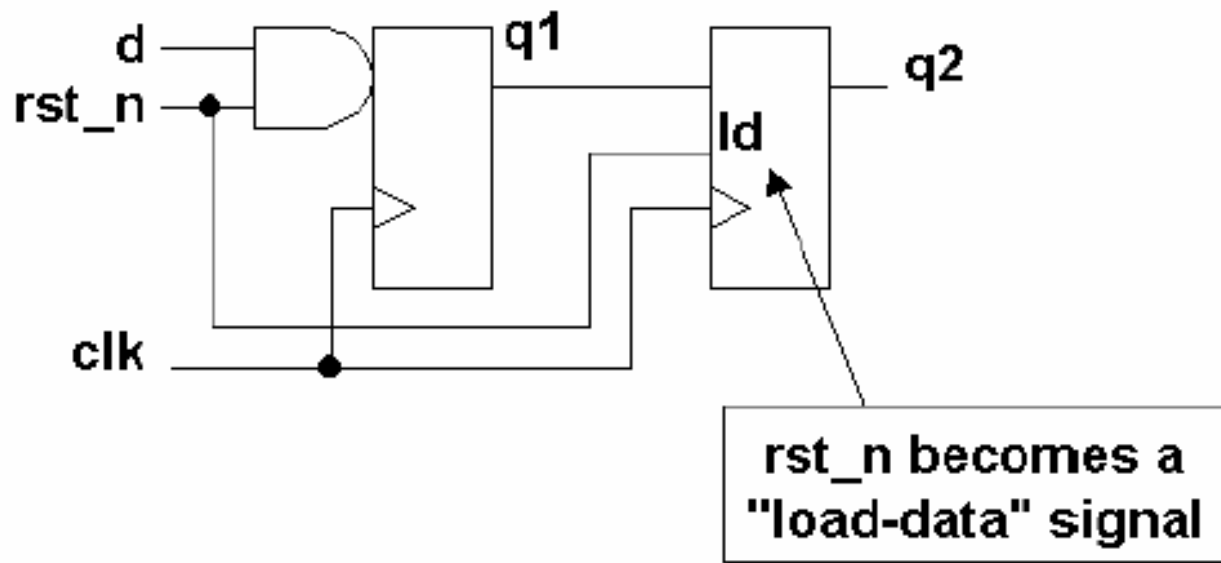
- Synchronous Reset:

```
module goodFFstyle (  
    output reg q2,  
    input d, clk, rst_n);  
    reg q1;  
    always @(posedge clk)  
        if (!rst_n) q1 <= 1'b0;  
        else q1 <= d;  
    always @(posedge clk)  
        q2 <= q1;  
endmodule
```

# Bad Reset

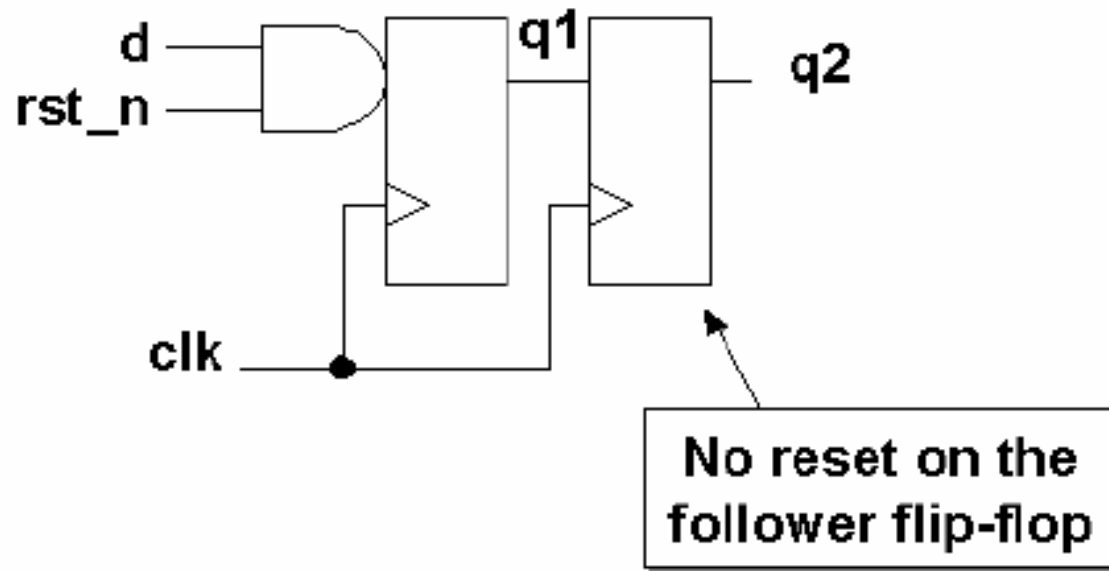
```
module badFFstyle (  
    output reg q2,  
    input d, clk, rst_n);  
    reg q1;  
    always @(posedge clk)  
        if (!rst_n) q1 <= 1'b0;  
        else begin  
            q1 <= d;  
            q2 <= q1;  
        end  
endmodule
```

# Bad Hardware



**Unnecessary use of a loadable flip flop**

# Good Hardware



**This is one of the few cases where a multiple always block is advised.**

# Pros of Synchronous Resets

- Flip flop size is less. Although the gate count increases.
- Circuit is 100% synchronous.
- Sometimes the reset may be an internally generated signal and may have glitches.
- Work as a filter for such reset glitches.
- However there can be a metastability if the glitches occur near the clock edges.

# Cons

- Not all ASIC libraries do have them
- Need a pulse stretcher to ensure that reset stays when the clock goes high
- Simulation issues can creep in, due to x-logic.
- Its often an annoying fact that you can do reset only when there is a clock. What if the clock is disable, say to save power?
- And would like to start the block.

# Good Reset Again!

- Asynchronous Reset:

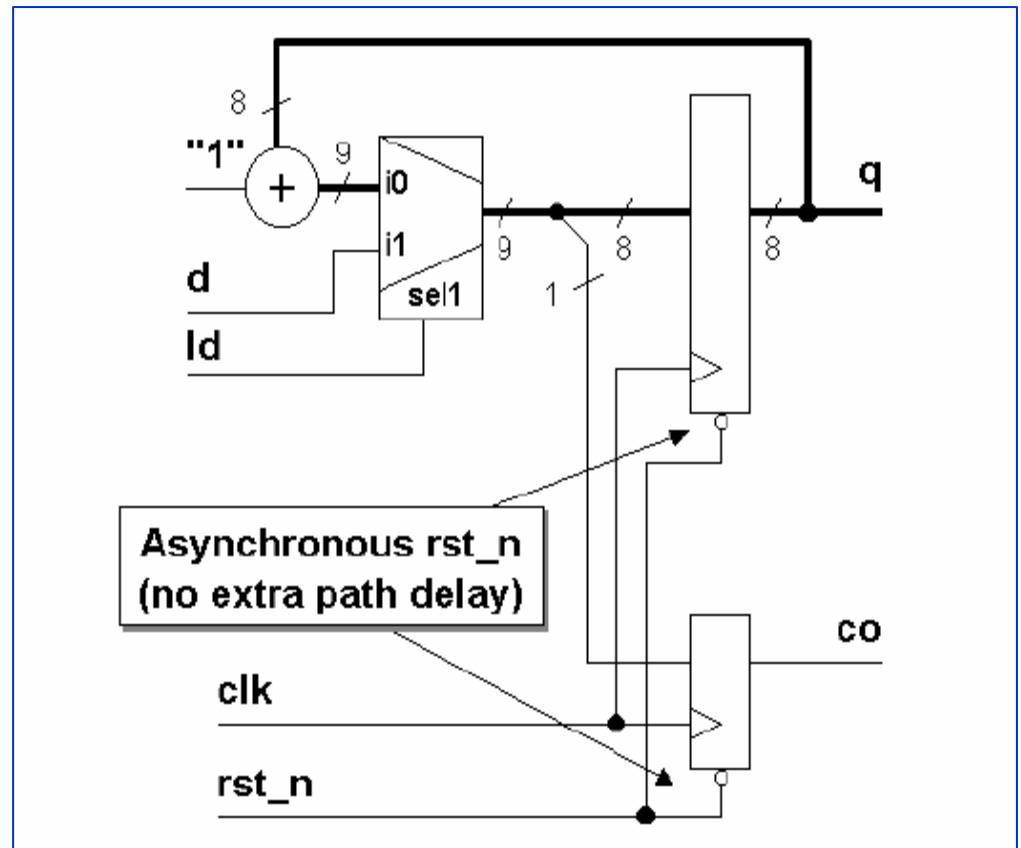
```
module good_async_resetFFstyle (  
    output reg q,  
    input d, clk, rst_n);  
    always @(posedge clk or negedge rst_n)  
        if (!rst_n) q <= 1'b0;  
        else q <= d;  
endmodule
```



# Pros

- Data Path is clean.  
Less gates on the data path.

```
module ctr8ar (  
  output reg [7:0] q,  
  output reg co;  
  input [7:0] d;  
  input ld, rst_n, clk;  
  always @(posedge clk or negedge  
    rst_n)  
    if (!rst_n) {co,q} <= 9'b0; // async  
                        //reset  
    else if (ld) {co,q} <= d; // sync load  
    else {co,q} <= q + 1'b1; // sync  
                        increment  
endmodule
```



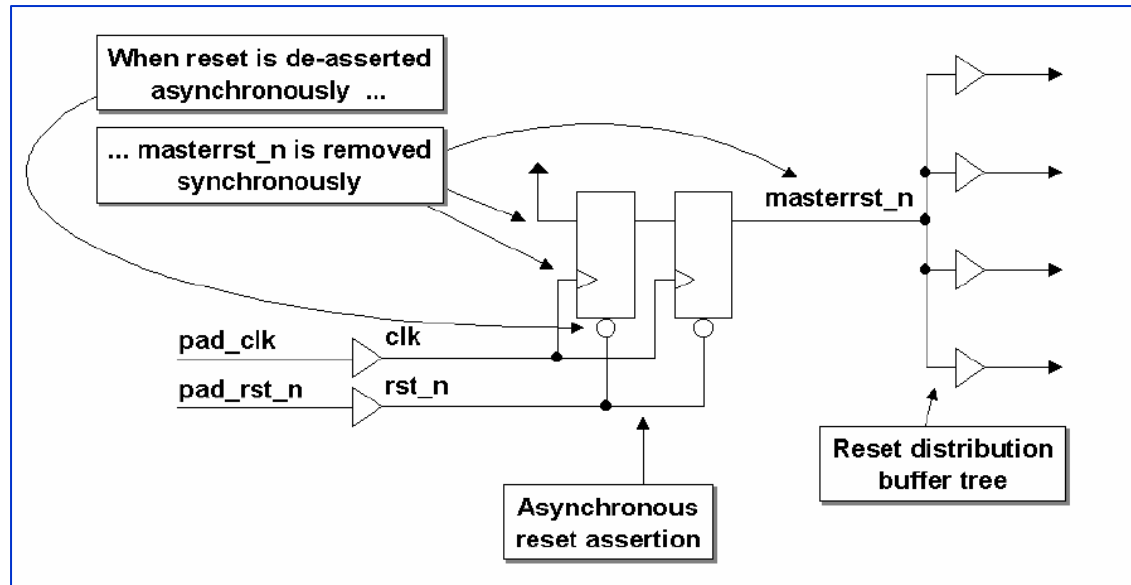
# Cons

- Difficult static Timing Analysis is hard.
- If the reset net is not derived from the input, they have to disable for scan based test (DFT issues).
- Glitches in the reset signal can be a problem.
- De-assertion of the reset could be an issue. If it happens near the active clock edge meta-stability can occur.

# Two main Cons...

- **Reset Recovery Time:** Time between the de-assertion of reset and the next active clock edge. If this is not obeyed, meta-stability can occur.
- **Reset removal is asynchronous:** Consider, the reset going to more than one flop. Due to the different propagation time in either or both the reset and the clock signal, some flops may be in reset state, while others may have gone passed the reset state.

# And Ugly Reset



- Two flip-flops are required to synchronize the reset signal to the clock pulse
- The second flip-flop is used to remove any metastability that might be caused by the reset signal being removed asynchronously and too close to the rising clock edge.
- You also have the best of asynchronous reset.
- Only reset becomes ugly!

# Reset using the reset synchronizer

```
module async_resetFFstyle2 (  
  output reg rst_n,  
  input clk, asyncrst_n);  
  reg rff1;  
  always @(posedge clk or negedge asyncrst_n)  
    if (!asyncrst_n) {rst_n,rff1} <= 2'b0;  
    else {rst_n,rff1} <= {rff1,1'b1};  
endmodule
```

# Conclusion

- We have seen various kinds of resets.
- Resets that are good, bad and ugly.
- One thing is clear reset is not simple. They should be carefully handled.